

## Solutions

2.1 a) Show that DTFT of  $x[n] * y[n]$  is  $X(e^{j\omega})Y(e^{j\omega})$

Starting w/ Definition

$$\text{DTFT} [x[n] * y[n]] = \sum_{n=-\infty}^{\infty} (x[n] * y[n]) e^{-j\omega n}$$

$$= \sum_{n=-\infty}^{\infty} \sum_{k=-\infty}^{\infty} x[k] y[n-k] e^{-j\omega n}$$

$$= \sum_{k=-\infty}^{\infty} x[k] \sum_{n=-\infty}^{\infty} y[n-k] e^{-j\omega n}$$

Let  $m = n - k$

$$= \sum_{k=-\infty}^{\infty} x[k] \sum_{m=-\infty}^{\infty} y[m] e^{-j\omega(m+k)}$$

$$= \sum_{k=-\infty}^{\infty} x[k] e^{-j\omega k} \sum_{m=-\infty}^{\infty} y[m] e^{-j\omega m}$$

$$= X(e^{j\omega}) Y(e^{j\omega})$$

b) Suppose  $x(t)$  is real. What can you conclude about FT?

Recall:

$$X(j\Omega) = \int_{-\infty}^{\infty} x(t) e^{-j\Omega t} dt$$

thus

$$X^*(j\Omega) = \int_{-\infty}^{\infty} x(t) e^{j\Omega t} dt \text{ if } x(t) \text{ is real}$$

$$X^*(j\Omega) = X(-j\Omega)$$

thus,

1) real part of  $X(j\Omega)$  is even

$$X_R(j\Omega) = X_R(-j\Omega)$$

2) Imaginary part is odd

$$X_I(j\Omega) = -X_I(-j\Omega)$$

3) Magnitude is even

$$|X(j\Omega)| = |X(-j\Omega)|$$

4) Phase is odd

$$\angle X(j\Omega) = -\angle X(-j\Omega)$$

C) Prove  $w(t) * (x(t) + y(t)) = w(t) * x(t) + w(t) * y(t)$

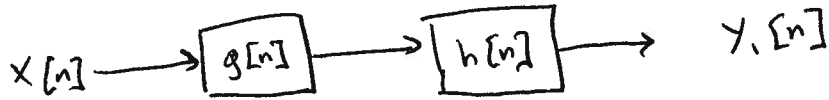
From definition

$$w(t) * (x(t) + y(t)) = \int_{-\infty}^{\infty} w(\tau) (x(t-\tau) + y(t-\tau)) d\tau$$

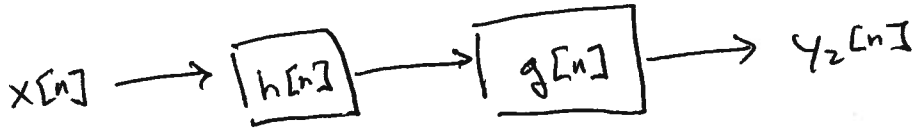
$$= \int_{-\infty}^{\infty} w(\tau) x(t-\tau) d\tau + \int_{-\infty}^{\infty} w(\tau) y(t-\tau) d\tau$$

$$= w(t) * x(t) + w(t) * y(t)$$

d) Show



and



then

$$y_1[n] = y_2[n]$$

2 Ways

- From part a) write

$$\begin{aligned}
 Y_1(e^{j\omega}) &= X(e^{j\omega}) G(e^{j\omega}) H(e^{j\omega}) \quad \text{clearly} \\
 &= X(e^{j\omega}) H(e^{j\omega}) G(e^{j\omega}) \\
 &= Y_2(e^{j\omega})
 \end{aligned}$$

so

$$y_1[n] = y_2[n]$$

- Or, write out convolution.

2.2

Consider continuous time signal

$$x(t) = \frac{\sin \pi t}{\pi t} \cdot \sin 9\pi t$$

Sketch the FT of  $X_s(\Omega)$  of  $x_s(t) = \sum_{n=-\infty}^{\infty} x(nT) \delta(t-nT)$

and the DTFT  $X(e^{j\omega})$  of  $x[n] = x(nT)$  for:

a)  $T = 1/14$

First, just sketch FT of  $x(t)$ .

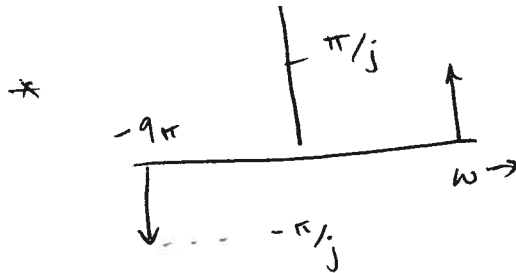
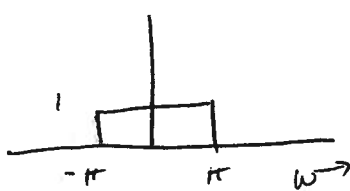
$$\text{so FT} \left( \frac{\sin \pi t}{\pi t} \right) = \begin{cases} 1 & |\Omega| \leq \pi \\ 0 & \text{else} \end{cases}$$

$$\text{FT}(\sin 9\pi t) = \frac{\pi}{j} \delta(\Omega - 9\pi) - \frac{\pi}{j} \delta(\Omega + 9\pi)$$

So  $X(j\Omega) = \frac{1}{2\pi} \text{FT} \left( \frac{\sin \pi t}{\pi t} \right) * \text{FT}(\sin 9\pi t)$

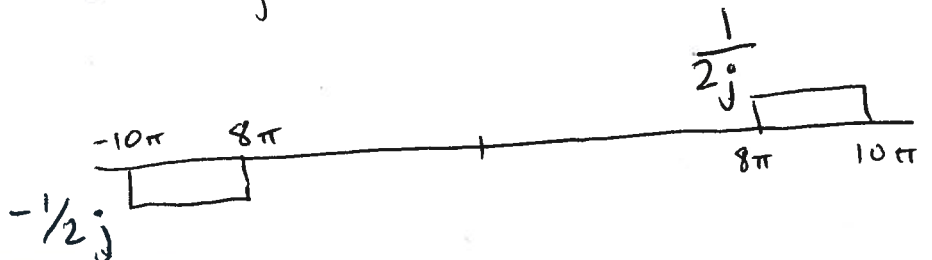
$$= \begin{cases} \frac{1}{2j} & 8\pi \leq \Omega \leq 10\pi \\ -\frac{1}{2j} & -10\pi \leq \Omega \leq -8\pi \\ \text{else} & \end{cases}$$

pic



$$\frac{1}{2\pi}$$

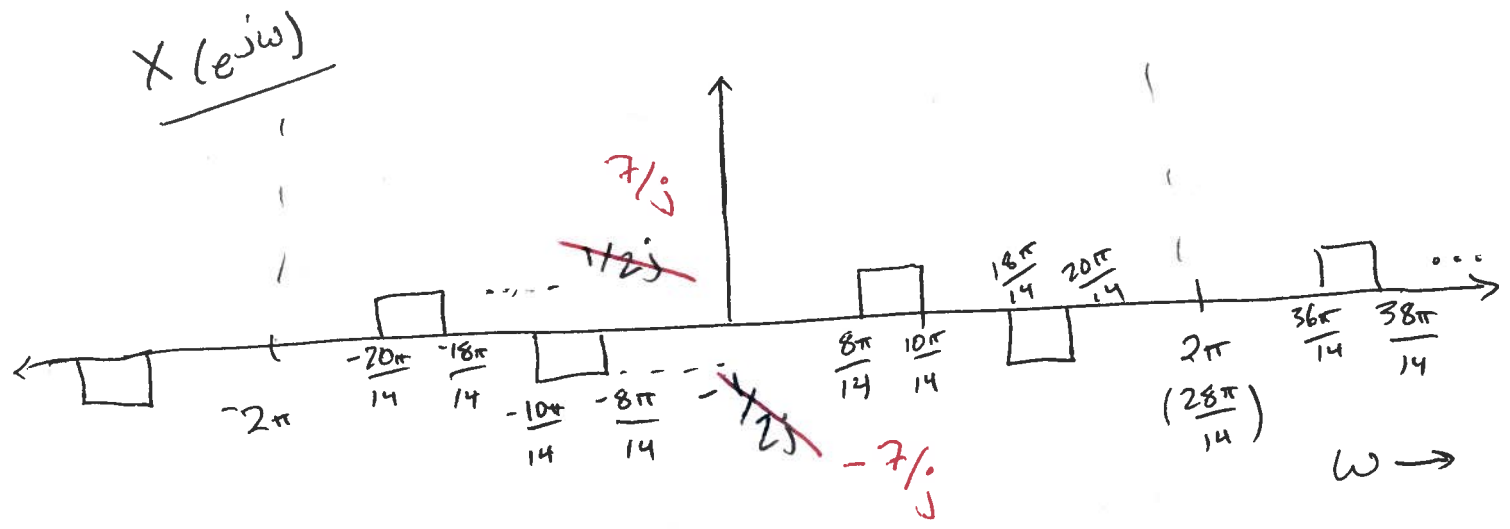
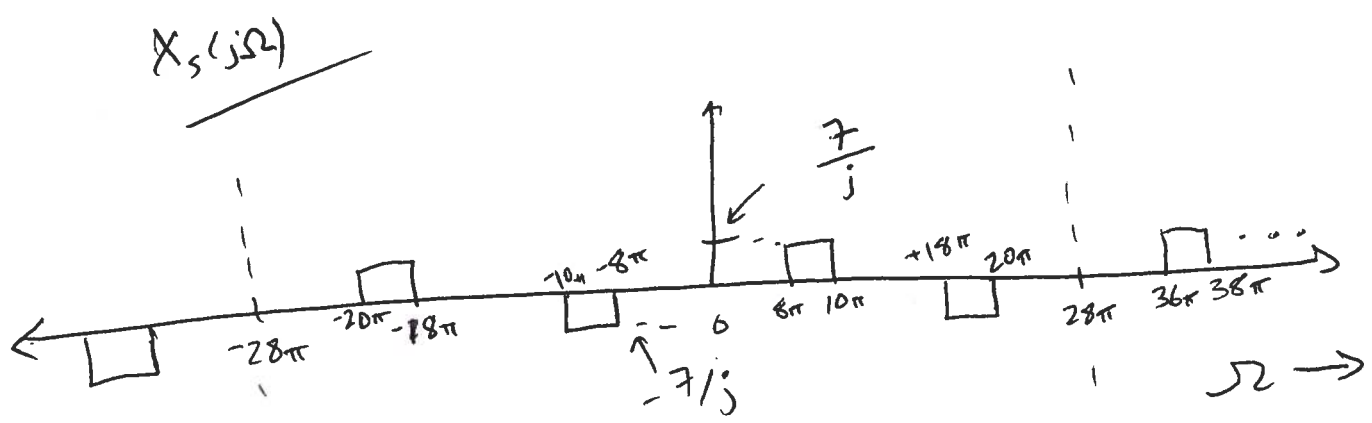
$$X(j\Omega) =$$



a/ continued

$$X_s(j\Omega) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X(j(\Omega - k\Omega_s))$$

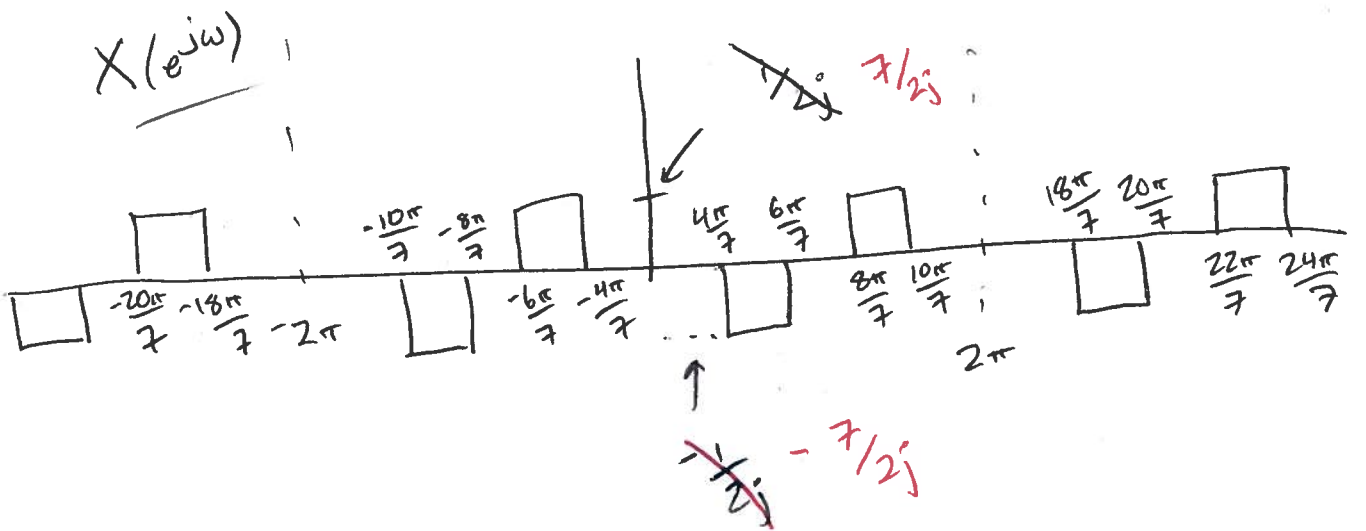
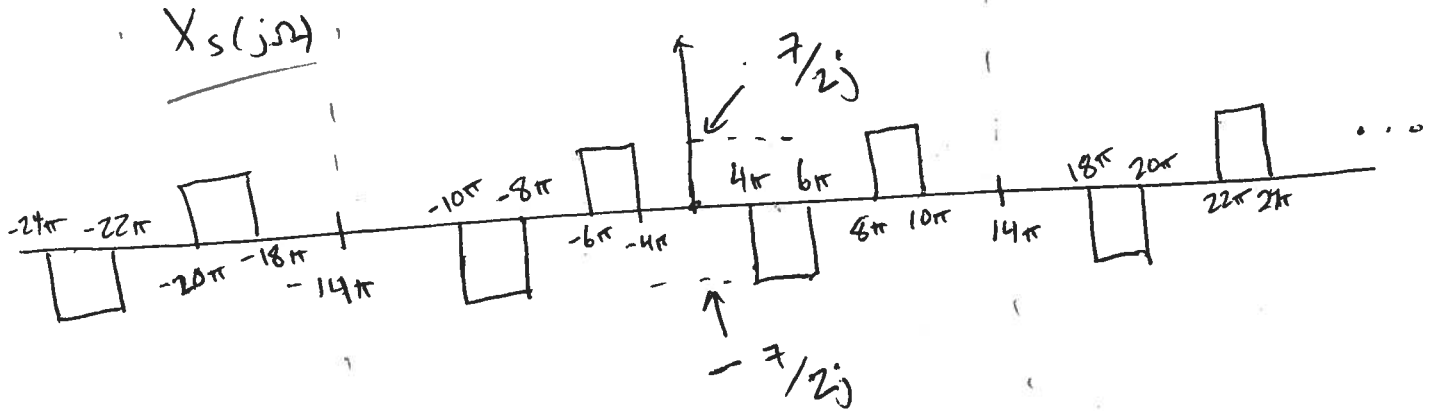
$\Omega_s = \frac{2\pi}{T} = 28\pi \rightarrow$  so, just have copies of  $X(j\Omega)$  every  $28\pi$ .



No Aliasing

5b

$$T = 1/7 \Rightarrow \Omega_s = \frac{2\pi}{T} = 14\pi$$

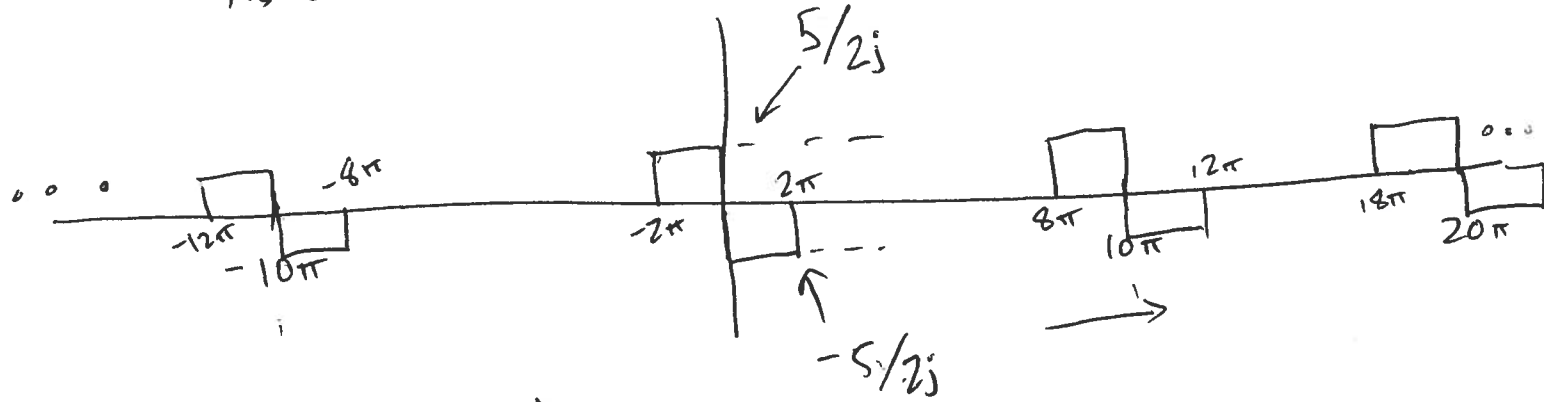


Yes aliasing occurs - but signals don't actually overlap, so we could recover continuous time signal with a-priori knowledge.

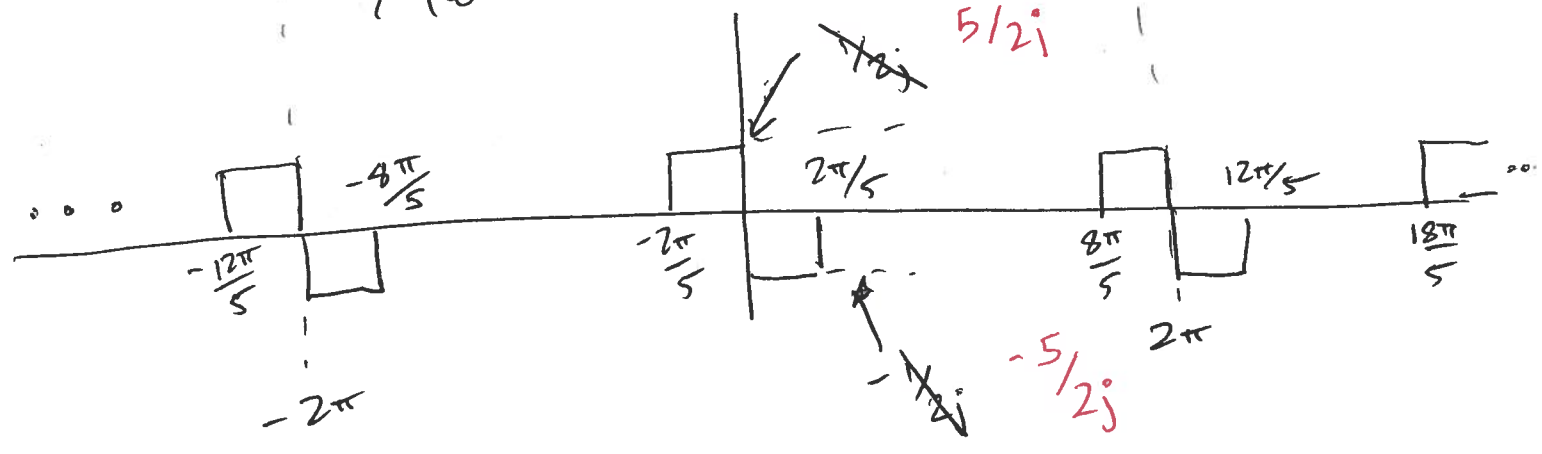
$$c/T = 1/5 \Rightarrow s = 10\pi$$

(7)

$$X_s(j\Omega)$$

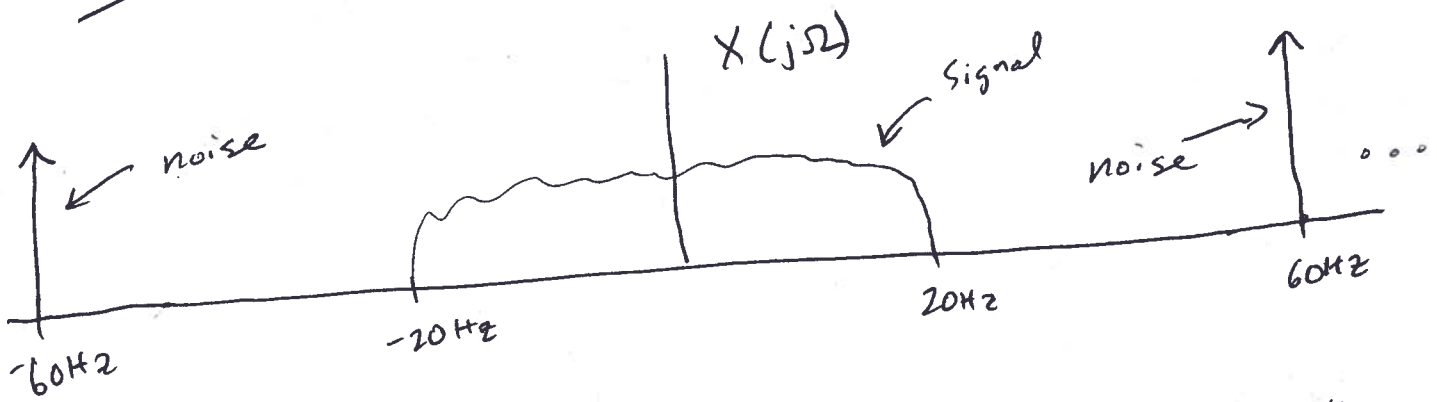


$$X(e^{j\omega})$$



Again, aliasing occurs, but we could completely recover C.t. signal with some a-priori knowledge.

2.3 EKG → first, draw signal in freq. (8)



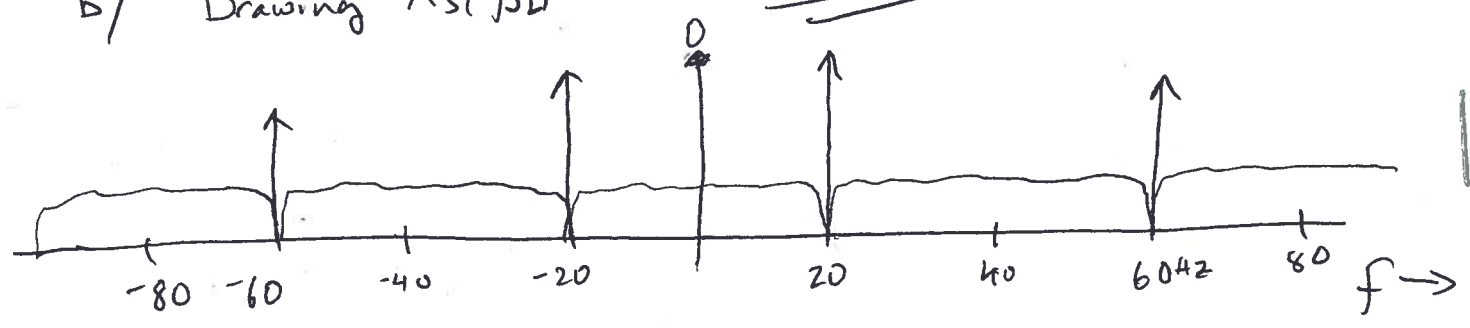
a) since our max signal frequency is 60 Hz  
(using notation from book)

$$\Omega_N = 2\pi f = 120\pi \rightarrow \Omega_s \geq 2\Omega_N = 240\pi$$

So,  $\Omega_s$  must be greater than  $\boxed{240\pi}$ .

Also, just can say we must sample at twice max frequency, so  $2 \cdot f_{max} = \underline{120 \text{ samples per sec.}}$

b/ Drawing  $X_s(j\Omega)$  2 answers

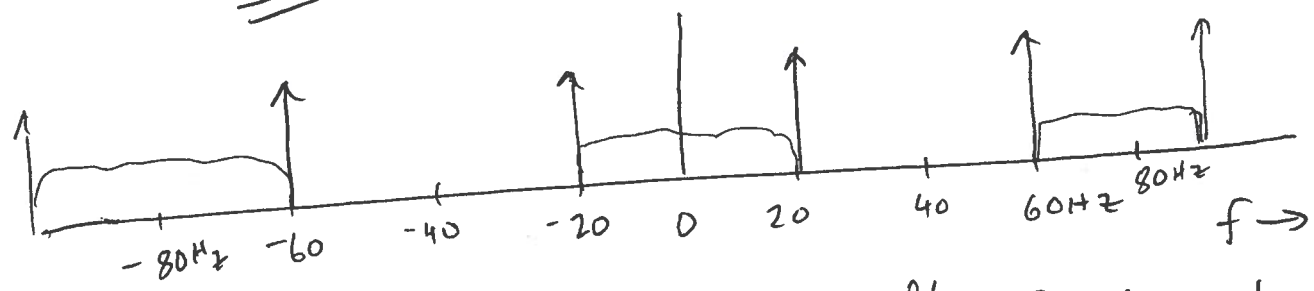


From picture, could sample at just 40 samples per second, or  $\boxed{80\pi = \Omega_s}$  if we can tolerate losing EKG signal right at 20 Hz. (Next)



2.3 continued

b/ OR  $X_s(j\Omega)$

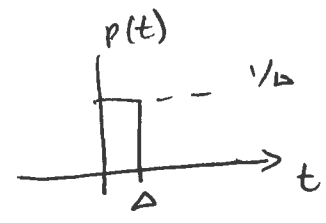


If we need signal at exactly 20 Hz also,  
 then, we must sample at  $> 80$  samples per  
 second, or  $\boxed{\Omega_s = 160\pi}$

→ we can tolerate aliasing since we know  
 there are open frequencies over our bandwidth.

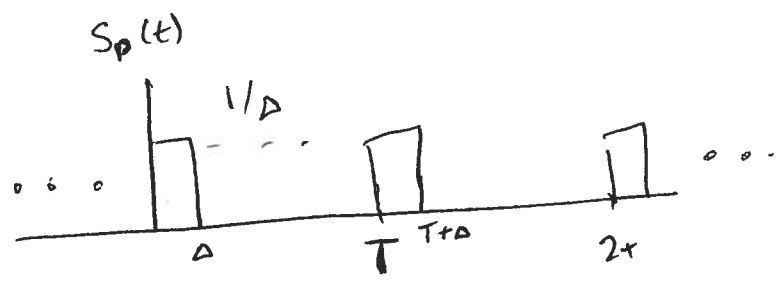
2.4

$$p(t) = \begin{cases} 0 & t < 0 \\ 1/\Delta & 0 < t < \Delta \\ 0 & t > \Delta \end{cases}$$



T > Δ

$$X_p = x(t) S_p(t) = x(t) \sum_{n=-\infty}^{\infty} p(t-nT)$$



So,  $S_p(t) = p(t) * \sum_{n=-\infty}^{\infty} \delta(t-nT)$

Thus  $X_p(t) = x(t) \cdot p(t) * \sum_{n=-\infty}^{\infty} \delta(t-nT)$

$$X_p(\Omega) = \frac{1}{2\pi} X(\Omega) * \left[ P(\Omega) \cdot \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(\Omega - k\Omega_s) \right]$$

$$P(\Omega) = \frac{2 \sin \Omega \Delta / 2}{\Delta \Omega} e^{-j\Omega \Delta / 2}$$

So finally

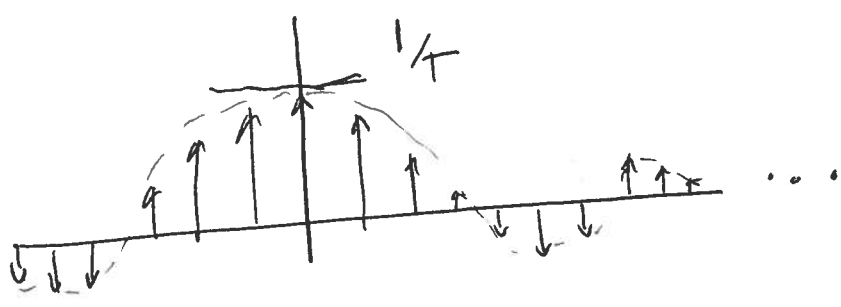
$$X_p(\Omega) = \frac{X(\Omega)}{2\pi} * \underbrace{\frac{2 \sin \Omega \Delta / 2}{\Delta \Omega} e^{-j\Omega \Delta / 2} \cdot \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(\Omega - k\frac{2\pi}{T})}_{\text{Sampled Sinc fun}}$$

(OVER)

Picture

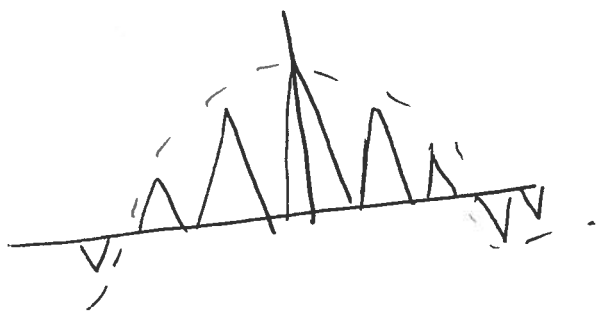
$$X_p(\Omega) = X(\Omega) * S_p(\Omega) = X(\Omega) *$$

$$S_p(\Omega) \text{ (sampled sinc fn)} \quad \frac{1}{T} \frac{2 \sin \Omega T/2}{\Delta \Omega} \sum_{k=-\infty}^{\infty} \delta(\Omega - k \frac{2\pi}{T})$$



Then, if  $X(\Omega)$  is band limited

$X_p(\Omega)$



and we can recover signal.