#### Solutions

### ECE431 Homework 1

Thursday, September 10, 2009

1. (OSB 2.30)

(a) Determine and sketch w[n] if  $x[n] = (-1)^n u[n]$ . Also, determine the overall output y[n].

**Solution:** The convolutions are most easily computed by drawing out the graphs. We can also compute the output directly from the equarrays. We write the first impulse response as

$$h_1[n] = u[n] - u[n-4].$$

The output after the first stage will be

$$w[n] = h_1[n] * x[n]$$
  

$$= \sum_{k=-\infty}^{\infty} h_1[k]x[n-k]$$
  

$$= \sum_{k=-\infty}^{\infty} (u[k] - u[k-4]) (-1)^{n-k}u[n-k]$$
  

$$= \sum_{k=-\infty}^{n} (u[k] - u[k-4]) (-1)^{n-k}$$
  

$$= \sum_{k=0}^{\min(n,3)} (-1)^{n-k}$$
  

$$= \begin{cases} 1 & n = 0, 2 \\ 0 & \text{else} \end{cases}$$
  

$$= \delta(n) + \delta(n-2)$$

The impulse response of the second system is

$$h_2[n] = u[n+3] - u[n-1].$$

We can write the output of the second stage as

$$y[n] = h_2[n] * w[n]$$
  
=  $(u[n+3] - u[n-1]) * (\delta(n) + \delta(n-2))$   
=  $(u[n+3] - u[n-1]) * \delta(n) + (u[n+3] - u[n-1]) * \delta(n-2)$   
=  $u[n+3] - u[n-1] + u[n+1] - u[n-3])$ 

$$= \left\{ \begin{array}{cc} 1 & n = -3, -2, 1, 2\\ 2 & n = -1, 0\\ 0 & \text{else} \end{array} \right\}$$

(b) Determine and sketch the overall impulse response of the cascade system; i.e. plot the output y[n] = h[n] when  $x[n] = \delta[n]$ .

**Solution:** We have

$$h[n] = h_1[n] * h_2[n]$$
  
=  $(u[n] - u[n-4]) * (u[n+3] - u[n-1])$   
=  $u[n] * u[n+3] - u[n] * u[n-1] - u[n-4] * u[n+3] + u[n-4] * u[n-1]$   
=  $nu[n+3] - nu[n-1] - nu[n-1] + nu[n-5]$   
=  $n(u[n+3] - 2u[n-1] + u[n-5])$ 



which can also be computed by graphing.

(c) Now consider the input  $x[n] = 2\delta[n] + 4\delta[n-4] - 2\delta[n-12]$ . Sketch w[n]. **Solution:** Again, we can write the output of the system as

$$w[n] = h_1[n] * x[n]$$
  
=  $h_1[n] * (2\delta[n] + 4\delta[n-4] - 2\delta[n-12])$   
=  $2h_1[n] - 4h_1[n-4] - 2h_1[n-12]$ 



Code to convolve two sequences in matlab

```
N = 40
n = [-N/2:1:N/2];
h1 = [zeros(1,20) 1 1 1 1 zeros(1,17)];
h2 = [zeros(1,17) 1 1 1 1 zeros(1,20)];
x = (-1).^n.*heaviside(n+.01);
%convolve h1 and h2 to get overall impulse response
for m = 1:1:N+1
    w(m)=x*flipud(circshift(h1.',-n(m)));
    %the above line uses x'x which multiplies and sums all elements of a
    %vector
end
w(1:10)=0;
w(length(w)-10:length(w))=0;
%zero padding
```

2. Consider an LTI system whose impulse response is  $h[n] = (0.5)^n u[n]$ .

(a) Determine the frequency response of the system. Sketch the magnitude of the frequency response. Is the system highpass, bandpass, or lowpass?

Solution: Starting with the definition of the discrete time Fourier transform, we have

$$H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} h[n]e^{-j\omega n}$$
$$= \sum_{n=-\infty}^{\infty} (0.5)^n u[n]e^{-j\omega n}$$
$$= \sum_{n=0}^{\infty} (0.5)^n e^{-j\omega n}$$
$$= \sum_{n=0}^{\infty} (0.5e^{-j\omega})^n$$
$$= \frac{1}{1 - 0.5e^{-j\omega}}$$



The system is low pass.

(b) Let  $x[n] = \cos(\pi n) + 4\cos(\frac{\pi}{2}n)$  be the input to this system. Compute the output y[n] = x[n] \* h[n].

Solution: We can compute the output of the system in the frequency domain. First,

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$
  
=  $\pi\delta(\omega - \pi) + \pi\delta(\omega + \pi) + 4\pi\delta(\omega - \frac{\pi}{2}) + 4\pi\delta(\omega + \frac{\pi}{2})$   $(-\pi \le \omega \le \pi)$ 

where  $X(e^{j\omega})$  is periodic with period  $2\pi$ . The output of the system is computed as

$$\begin{split} Y(e^{j\omega}) &= H(e^{j\omega})X(e^{j\omega}) \\ &= \frac{1}{1 - 0.5e^{-j\omega}} \left( \pi \delta(\omega - \pi) + \pi \delta(\omega + \pi) + 4\pi \delta(\omega - \frac{\pi}{2}) + 4\pi \delta(\omega + \frac{\pi}{2}) \right) \\ &= \frac{\pi}{1 - 0.5e^{-j\pi}} \delta(\omega - \pi) + \frac{\pi}{1 - 0.5e^{j\pi}} \delta(\omega + \pi) + \frac{4\pi}{1 - 0.5e^{-j\pi/2}} \delta(\omega - \frac{\pi}{2}) + \frac{4\pi}{1 - 0.5e^{j\pi/2}} \delta(\omega + \frac{\pi}{2}) \quad (-\pi \le \omega \le \pi) \end{split}$$

Transforming the above equation back to the time domain, we have

$$y[n] = \frac{2}{3}\cos(\pi n) + \frac{16}{5}\cos(\frac{\pi}{2}n) + \frac{8}{5}\sin(\frac{\pi}{2}n)$$

This agrees with our intuition - when we input pure sinusoids into an LTI system, the output is a pure sinusoids of the same frequency.

(c) Let  $x[n] = \delta[n-1] + 3\delta[n-3]$  be the input to the system. Compute the output y[n] = x[n] \* h[n].

**Solution:** Unlike part (b), the output is most easily computed in the time domain.

$$y[n] = x[n] * h[n]$$
  
=  $(\delta[n-1] + 3\delta[n-3]) * 0.5^{n}u[n]$   
=  $0.5^{n-1}u[n-1]0.5^{n-3}u[n-3]$   
=  $0.5^{2n-4}u[n-3]$ 

3. Download the MATLAB data file mclips.mat and the MATLAB program music.m from the web site. The program will playback either of the two music clips x and y contained in mclips.mat. The clip y is a version of x that has been digitally processed in MATLAB to produce an echo effect.

(a) Design and implement a linear system in MATLAB that processes the original clip x to produce an echo with a 0.1 second delay (this should sound similar to y). Turn in the derivation of your echo system and the MATLAB code that implements it.

(b) Design and implement a nonlinear system in MATLAB that produces a saturation effect. That is, mimic the effect of pushing an amplifier into saturation by clipping x whenever the magnitude of its amplitude is greater than a certain level (e.g., |x[n]| > 0.7). Turn in your design and the MATLAB code that implements it.

Solution: Generated using MATLAB publishing -

## 0.1 Homework 1 Problem 3

close all
clc
clear
load mclips.mat
% plot the audio signal x
plot(x)

# 0.2 Part a) Create Echo

yecho = [x; zeros(fs,1)]+ [zeros(fs,1); x]; yecho = yecho/max(yecho);

## 0.3 Part b) Clip Audio

yclip = y; yclip(yclip>0.3) = 0.3; yclip(yclip<-0.3) = -0.3; sound(yclip,fs) plot(yclip) axis([0 length(yclip) -1 1])

4. Verify the inverse DTFT formula

$$h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(\omega) e^{j\omega n} d\omega$$

by replacing  $H(\omega)$  with the DTFT expression

$$H(\omega) = \sum_{k=-\infty}^{\infty} h[k]e^{-j\omega k}$$

and show that you obtain h[n] on both sides of the equarray.

Solution: As the problem suggests, combine the two equarrays.

$$h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} \sum_{k=-\infty}^{\infty} h[k] e^{-j\omega k} e^{j\omega n} d\omega$$
$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} \int_{-\pi}^{\pi} h[k] e^{-j\omega(k-n)} d\omega$$
$$= \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} h[k] \int_{-\pi}^{\pi} e^{-j\omega(k-n)} d\omega$$

Evaluating the integral, and using Euler's Identity, we have

$$\int_{-\pi}^{\pi} e^{-j\omega(k-n)} d\omega = \frac{e^{-j\pi(k-n)}}{-j(k-n)} - \frac{e^{j\pi(k-n)}}{-j(k-n)} = 2\frac{\sin(\pi(k-n))}{(k-n)}$$

Notice that  $2\sin(\pi(k-n))/(k-n)$  is zero for all integer values of (k-n), except when (k-n) = 0. In this case, since both the numerator and denominator go to zero, we define the answer to be the limit as (k-n) goes to zero. Using l'Hopital's rule

$$\lim_{(k-n)\to 0} \frac{2\sin(\pi(k-n))}{(k-n)} = 2\pi$$

and thus

$$2\frac{\sin(\pi(k-n))}{(k-n)} = 2\pi\delta(k-n).$$

Finally, we have

$$h[n] = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} 2\pi h[k]\delta(k-n)$$

All the terms in the sum are zero unless n = k, and we have

$$h[n] = \sum_{k=n}^{n} h[k] = h[n]$$